Development and deployment of large scale wireless sensor network on a long-span bridge

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Abstract. Testing and validation processes are critical tasks in developing a new hardware platform based on a new technology. This paper describes a series of experiments to evaluate the performance of a newly developed MEMS-based wireless sensor node as part of a wireless sensor network (WSN). The sensor node consists of a sensor board with four accelerometers, a thermometer and filtering and digitization units, and a MICAz mote for control, local computation and communication. The experiments include calibration and linearity tests for all sensor channels on the sensor boards, dynamic range tests to evaluate their performance when subjected to varying excitation, noise characteristic tests to quantify the noise floor of the sensor board, and temperature tests to study the behavior of the sensors under changing temperature profiles. The paper also describes a large-scale deployment of the WSN on a long-span suspension bridge, which lasted over three months and continuously collected ambient vibration and temperature data on the bridge. Statistical modal properties of a bridge tower are presented and compared with similar estimates from a previous deployment of sensors on the bridge and finite element models.

Keywords: wireless sensor network; testing and calibration; MEMS; modal analysis; bridge monitoring.

1. Introduction

Advances in micro-electro-mechanical-systems (MEMS) technology as well as wireless network systems have attracted attention to the applications of wireless sensor networks (WSNs) in different areas of engineering. One such area of interest is the implementation of WSNs for structural monitoring applications. Several research groups have developed hardware and software wireless sensor network (WSN) platforms to address different requirements of this application and/or deployed a prototype system on civil structures. Spencer *et al.* (2004), Spencer (2003), Ruiz-Sandoval *et al.* (2006), Ruiz-Sandoval *et al.* (2003), Lynch *et al.* (2004), Whelan and Janoyan (2009), Hackmann *et al.* (2005), Aoki *et al.* (2009), Cho *et al.* (2008) and Rice and Spencer (2008) are a few examples of such studies. A critical task in developing a new hardware platform which is based on advanced technology is the testing and validation processes. Only after the primary performance of the new systems have been validated can their advantages in terms of low-cost, ease of installation and maintenance, and superior computational capabilities be utilized.

Pakzad et al. (2008) describes an integrated platform for structural monitoring applications, which

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consists of a sensor board with four channels of accelerometers and a thermometer. The sensor board is equipped with dedicated filtering and digitization chips. This platform accepts MICAz motes for control and communication and runs a control software based on TinyOS operating system (Hill et al. 2003). Building up on that work, this paper presents testing and validation processes for the hardware through laboratory and field experiments. The objective of this paper is to present several testing protocols that can be used to evaluate the performance of the hardware for various circumstances. Several laboratory tests are described which examine the behavior of the sensor board under static and dynamic excitations and quantify their characteristic noise levels. Calibration parameters for each sensor are estimated and linearity of their response is evaluated. The behavior of the sensor board under varying temperature profiles is tested and response histograms are presented. The integrated system is deployed on a long-span bridge and the results of statistical modal identification of a tower of the bridge are presented and compared with similar estimates from a previous deployment of sensors on the bridge and finite elements models.

The organization of the remainder of the paper is in three main parts. The hardware platform and its different components are described in section 2. Then the laboratory testing and calibration processes are presented in section 3. Sections 4 and 5 describe the field deployment of the WSN on Golden Gate Bridge as a testbed and present the analysis results.

2. Hardware platform

The wireless sensor network consists of a set of connected sensor nodes. Each node has three main hardware components: sensors, filters and micro-controller, and radio for wireless communication. The sensor board, which has the sensing, filtering and analog-to-digital converters, is shown in Fig. 1. These hardware components are discussed in this section in more details.

2.1 Sensors

The main sensors on the node are MEMS accelerometers. For the measurement of low-level and high-level accelerations, two commercially available sensor units are used in each of two directions

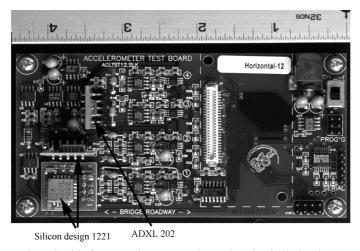


Fig. 1 Sensor board with four accelerometer channels (physical size is 115 × 75 mm)



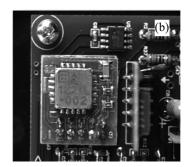


Fig. 2 (a) ADXL 202 sensor (physical size $5 \times 5 \times 2$ mm) and (b) Silicon Designs 1221 sensor installed on the sensor board (physical size $3.5 \times 3.5 \times 1$ mm)

(one horizontal and one vertical). The use of two sensor types is a cost-effective solution and allows examination of performance-price tradeoffs. The high-level sensor is Analog Device's ADXL 202 shown in Fig. 2(a), a widely used device that provides $\pm 2.0~g$ range with a sensitivity of 1 mg at 25 Hz (Analog Devices 1999). It is a low-cost (~\$10 in 2005), low-power complete two-axis accelerometer, which can measure both dynamic acceleration (vibration) and static acceleration (gravity). Its dimensions are 5 mm \times 5 mm \times 2 mm. Its nominal resolution is 200 μ g per root-square Hz, allowing signals below 2 mg (at 60 Hz bandwidth) to be resolved. Accelerometer noise has the characteristic of white Gaussian noise that contributes equally at all frequencies. This accelerometer provides the required range of strong motions for the node.

Fig. 2(b) shows a Silicon Designs 1221, which provides acceptable sensitivity for low-level ambient structural vibrations at a relatively low cost (Silicon Designs 2007). This single-axis sensor also responds to both static and dynamic vibrations, provides the sensitivity required for ambient vibrations, and has a nominal noise level of 2 μ g per root-square Hz. The range of Silicon Designs device is reduced to ± 0.150 g so by using 16-bit A/D converters maximum nominal resolution of 4.6 μ g is achieved. Tests show that the noise floor of the accelerometer itself is controlling and the device is sensitive to within 10 μ g (see the following section for more details).

The other sensor on the board is a thermometer to provide accurate temperature data at each node. This is an important piece of information since tests showed that the sensors are sensitive to temperature changes and the temperature information can be used to calibrate the sensors.

2.2 Filtering and analog/digital conversion

Each channel from the MEMS accelerometers provides an analog voltage that is fed to a single-pole (-6 db) anti-aliasing low-pass filter with a cutoff frequency of 25 Hz. The filter was set for a long-span bridge application because even very high vibration modes have frequencies well below the cutoff frequency. The filtered analog signal is fed to 16-bit analog-to-digital converter (ADC) for each of the four channels. Sampling is done at a high frequency (1 kHz), but the digitized signal is downsampled by averaging, which acts as a digital filter and reduces the Gaussian noise level by a factor of \sqrt{n} , when every n samples are averaged. Combination of the 25 Hz analog anti-aliasing filter, high-frequency sampling at 1000 Hz, and downsampling, provides a simple and power efficient approach for high-resolution acceleration measurement.

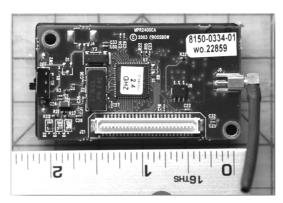


Fig. 3 MICAz Mote (physical size is 56×25 mm)

2.3 Control and communication unit

The mote with a micro-controller provides local processing and storage capability and a low-power radio communication for a node. The MICAz mote was selected because it has a good tradeoff between processing and communication capability, and power requirements (Crossbow Technology, 2007). The MICAz, pictured in Fig. 3, has 512 kB flash memory, which can store up to 250,000 two-byte data samples, and a 2.4 GHz radio-frequency (RF) Chipcon CC2420 transceiver with a hardware interface that can support commercially available bi-directional antennas. The ability of the mote to connect with a bi-directional antenna was an important factor because the long-span bridge application required a linear network topology and a standard omni-directional antenna would have wasted a significant amount of radio power. Based on the power usage testing of the node (Pakzad *et al.* 2008), it was decided to use four 6 V lantern batteries to provide 12 V and 15 Ah in the deployment.

3. Testing and validation

Micro-electro-mechanical-system (MEMS) devices are a relatively new category of integrated electronic chips and therefore do not have a long track record of performance for various applications. This makes the process of testing and validation of the hardware and control software critical, because one of the tasks for any research project that uses these devices is to evaluate and validate their performance. Only after establishing such performance measures, can the advantages of MEMS devices in terms of small size, low cost, ease of installation and maintenance, and eventually scalability be considered. In this section the testing and calibration of the sensor network for structural monitoring is presented. Four major categories of tests were performed on the accelerometer sensors: calibration and linearity tests, static tests, shaking table tests, and temperature sensitivity tests. The calibration parameters were estimated for use in further testing and comparison with other devices. The static tests evaluated the performance of the sensors under ambient conditions to establish the noise characteristics of the accelerometers. The shaking table tests were performed to analyze the dynamic range of the sensors and test the response characteristics in the frequency range of interest. The

temperature tests evaluated the sensitivity of the accelerometer sensors to temperature change. Each set of tests and results are presented in the following sections.

3.1 Calibration and linearity tests

The sensor boards were manufactured in two categories: *horizontal boards*, and *vertical boards*. The horizontal boards are designed for the bridge span to measure acceleration in transverse and vertical directions, so the range of the low-level Silicon Designs 1221 sensor in vertical direction is adjusted to gravity to include 850 mg to 1150 mg interval. The vertical boards, designed to measure accelerations in transverse and longitudinal directions, do not need the adjustment for the gravity and can be used on the bridge tower. Overall, seventy-eight horizontal boards and twenty vertical boards were manufactured.

The accelerometer sensors were calibrated so the digital output of the analog/digital converter (ADC) can be translated into an acceleration value. Calibration was performed by subjecting the sensor to known static accelerations and determining the digital output for that input. The calibration process is performed using the acceleration due to gravity by tilting the sensor between a position parallel to the ground that produces 1.0 g, and vertical position, which corresponds to zero acceleration. At each stop the ADC readings are recorded and compared with the corresponding acceleration and the linear shift and scale factors are estimated. Several stops between vertical and horizontal positions are used to examine the linearity of the accelerometer response. This method has the advantage of simplicity and ease of use, and does not require complicated equipment. It is, however, limited to ± 1.0 g range, so it is required to assume that the response remains linear for larger accelerations. Another implicit assumption is made that the acceleration readings are independent of frequency of the input signal and the DC calibration is valid for dynamically applied excitations.

For field deployment, the calibration parameters for every sensor on each board, i.e., shift and scale parameters, were individually determined with tilt tests. A Yuasa 220 single axis rotary tilt table with a UDNC-100 programmable controller was used to tilt the sensor boards, as shown in Fig. 4. The rotary table has a resolution of 0.001 degrees, with an optional horizontal or vertical setting position. The programmable controller was set up to rotate the board under test to a station with a specific angle and collect 1500 samples from each of the four accelerometer channels, before rotating the board further to the next station and repeating this process. The thirteen predefined stations and the corresponding angles and accelerations are listed in Table 1. The horizontal sensors

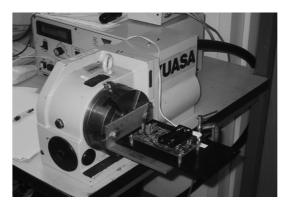


Fig. 4 The Yuasa 220 single axis rotary tilt table and UDNC-100 controller during the calibration of a sensor board

Station	Angle	Vertical acceleration (mg)	Horizontal acceleration (mg)
1	270	0.0	-1000.0
2	330	866.0	-500.0
3	345	965.9	-258.8
4	352	990.3	-139.2
5	354	994.5	-104.5
6	357	998.6	-52.3
7	0	1000.0	0.0
8	3	998.6	52.3
9	6	994.5	104.5
10	8	990.3	139.2
11	15	965.9	258.8
12	30	866.0	500.0
13	90	0.0	1000.0

Table 1 Tilt stations and their corresponding accelerations for calibration of vertical boards using Yuasa rotary table

were subjected to the full range of -1.0 g to +1.0 g with the rotation of one-half revolution about the horizontal axis.

The ADXL 202 sensor measures the whole interval since its operational range is ± 2.0 g. The Silicon Designs 1221 sensor saturates at Station 1, and remains so until Stations 4 or 5, where its operational range of approximately ± 150 mg is reached. This channel saturates again at Stations 9 or 10 where the applied acceleration exceeds its range. In the vertical direction the corresponding acceleration at Station 1 is zero, which is in range for the ADXL 202 but out of range for the Silicon Designs 1221. At Stations 2 through 12 the acceleration goes from 866 mg to 1.0 g and again back to 866 mg, which is measured by both vertical sensors. The Silicon Designs 1221 sensor saturates at Station 13. Fig. 5 shows the rotation angle θ with respect to the horizontal direction. The acceleration generated in vertical and horizontal sensors when the board is tilted by an angle θ is

$$\begin{cases} Acc_{Hor} = Sin(\theta) \cdot g \\ Acc_{Ver} = (1 - Cos(\theta)) \cdot g \end{cases}$$

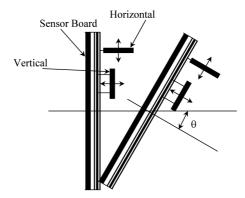


Fig. 5 Schematic tilt of sensor board by an angle θ

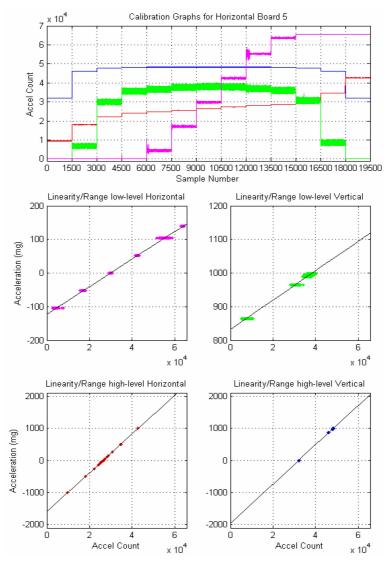


Fig. 6 Calibration plots for horizontal Board 5

Fig. 6 shows typical calibration results for one of the horizontal boards. The figure also shows the linearity and range of the sensors under test. The linearly fit line extends from zero count to the full count (65535) to extrapolate the actual range of the sensor. This line varries slightly from sensor to sensor, and from board to board depending on manufacturing and circuitry differences.

The data confirm that all four accelerometer channels on all sensor boards have a linear response to tilt, and hence shift and scale factors would be sufficient to calibrate the digital outputs.

3.2 Static noise characteristics tests

Every sensor has a characteristic noise level which represents sensor circuitry's noise floor under no external excitation. The purpose of static noise tests is to estimate this noise floor and compare the performance of the sensors with other devices with known noise characteristic in a quiet environment.

The static tests consist of two experiments that were conducted at Berkeley Seismological Laboratory facilities. The first test was performed in a laboratory in McCone Hall, and the other in a vault in Tilden Hill, near UC-Berkeley campus. The test in Berkeley Seismological Laboratory was performed in the second floor of McCone Hall, under ambient vibrations in an active environment of an academic building. The prototype sensor board was attached to the floor of the building and acceleration data at the rate of 580 Hz was collected for 1600 seconds. This test was then repeated in a seismic vault in Tilden Hill near Lawrence Berkeley Laboratories (LBL Vault), which is located inside a cave on solid rock, and houses Berkeley Seismological Laboratory's low-noise seismometers. A BKS FBA-23 reference accelerometer, with the root mean square (RMS) noise level of 0.2 μ g in the [0 25] Hz interval, was used as the reference sensor to compare the recorded vibrations and estimate the noise floors (results not shown here). Fig. 7(b) shows the prototype sensor board, weighted down by a heavy piece of lead inside the LBL Vault during the test.

The signals from the static noise tests in both time and frequency domains are presented in Fig. 7(a). The time histories are plotted for 1600 sec, and the power spectral densities are limited to 50 Hz. Both sets of plots confirm that the signal from McCone Lab is noisier than the one from LBL Vault. The sensor board experienced a noticeable temperature change during the LBL Vault

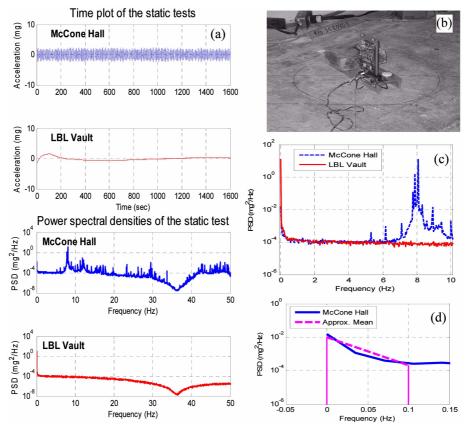


Fig. 7 (a) Time and Frequency plots for static test at McCone Lab and LBL Vault, (b) sensor board in LBL Vault, (c) zoom plot of the noise power spectral density of Silicon Designs 1221 sensor at LBL Vault and McCone Lab and (d) zoom plot of the very low noise power spectral density

test, which is the likely source of the small bump at the beginning of the time history plot and the slow drift afterwards. Otherwise, the signal remains between $\pm 100~\mu g$ of the de-trended level. The signal from the McCone Lab varies in an interval of $\pm 2~mg$, with a distinct frequency that can be observed in plots of both time and frequency domains. The likely source of the 8~Hz peak in the power spectral density (PSD) at McCone Lab is that this is a transverse vibration mode of the building, which has a very stiff reinforced concrete shear walls. A zoom plot of the power spectral densities of both signals is shown in Fig. 7(c). This graph shows that the low-frequency noise level of the sensor is at -40 dB, so for frequencies over 0.1 Hz, the root mean square acceleration noise per root-Hz is calculated as follows

$$e_{f>0.1Hz} = \sqrt{10^{-4}} = 10^{-2} \text{ mg}/\sqrt{Hz} = 10 \ \mu\text{g}/\sqrt{Hz}$$

There is, however, a significant increase in the noise level for frequencies lower than 0.1 Hz. Assuming -20 dB noise power at DC and -37 dB at 0.1 Hz (see Fig. 7(d)), the equivalent RMS noise level per root-Hz in the [0 0.1] Hz is estimated by

$$e_{f<0.1Hz} = \sqrt{\frac{10^{-2} + 2 \cdot 10^{-4}}{2}} = 0.071 \text{ mg}/\sqrt{Hz} = 71 \mu g/\sqrt{Hz}$$

The relatively flat part of the PSD for the tests at McCone Lab and LBL Vault in Fig. 7 is due to the characteristic noise of the accelerometer and amplifier, which shows that the self-noise of the Silicon Designs 1221 sensor is higher than the background low frequency noise at both test locations.

The digitization sensitivity of the Silicon Designs 1221 sensors after limiting the range to ± 150 mg and using a 16-bit ADC is 300 mg/2¹⁶-1=4.6 μ g, which is less than the characteristic self-noise of the sensors. These static tests confirm that the noise floor of the low-level high-sensitivity Silicon Designs 1221 accelerometers are well within the target noise level of less than 100 μ g for ambient vibration monitoring of a long-span bridge.

3.3 Dynamic range shaking table tests

The static tests of the prototype sensor boards provided valuable information about the noise characteristics of the sensors, but they do not say much about the response accuracy or range under dynamic excitation. The purpose of the shaking table tests, conducted during the development of the prototype boards, was to confirm that the sensors have predictable response when subjected to dynamic motion. To this end, a mass-spring low-noise short-period vertical shaking table constructed from a Johnson-Matheson Model 6840 was used to subject the sensors in the vertical direction to harmonic vibrations with varying amplitudes. Fig. 8(a) shows the set up of the tests. A reference Wilcoxon 731-4A low-noise piezoelectric accelerometer was also placed on the shaking table. A harmonic function generator was used to dynamically drive the table at frequencies 0.5, 1, 2, 4 and 5 Hz, each for approximately 30 sec. Both high-level ADXL 202 and low-level Silicon Designs 1221 sensors in vertical direction were used to collect data during the tests.

Fig. 8(b) shows the time histories of the recorded signals for the prototype sensor boards as well as the reference accelerometer. Power spectral densities using Welch method for the signals from three sensors are plotted in Fig. 8(c). Data from low-level Silicon Designs 1221 in both time and frequency domains are very similar to that for the reference low-noise sensor. The ADXL 202 sensor has a higher noise floor; it is especially evident that in the low-frequency area of the plot the

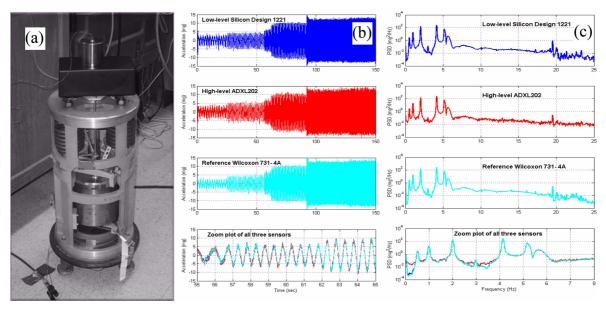


Fig. 8 (a) Sensor board and a piezoelectric reference accelerometer Wilcoxon 731-4A on the low-noise vertical mass-spring shaking table, (b) time history plots for the dynamic test of prototype Board 1 and (c) power spectral densities for the dynamic test of prototype Board 1

noise floor of the ADXL 202 is about 20 dB higher than Silicon Designs 1221. All sensors in both experiments show a higher DC noise level that particularly affects [0 0.1] Hz interval, undermining the advantage of using Silicon Designs 1221 in the very-low-frequency range. The similar time series of the dynamic tests and their corresponding power spectral densities demonstrate that the MEMS accelerometers accurately measure dynamic accelerations in the frequency range of interest.

3.4 Temperature sensitivity tests

MEMS accelerometers are temperature sensitive devices because the piezoelectric and flexibility properties of the silicon oscillators vary with temperature. The variation in the recorded acceleration during the static test on the LBL Vault, shown in Fig. 7(a), is an example of this temperature sensitivity. To understand this phenomenon a series of temperature tests were performed on the prototype sensor boards using an oven with controlled heating and cooling devices. Fig. 9(a) shows the temperature and acceleration recording of one typical test. The temperature time history plot represents the temperature profile of the test, which starts at room temperature. After an initial period of soaking, the oven cools at the rate of one degree per minute for five minutes, followed by a five-minute period of soaking until it reaches five degrees. After soaking at this temperature for five minutes, the warming period begins, again at a rate of one degree per minute for five minutes followed by soaking for five minutes, up to forty-five degrees, and then cooling again to twenty-five degrees. The second part of the profile consists of continuous warming/cooling, followed by longer soaking periods. The middle plot in Fig. 9(a) shows the acceleration response to this temperature change. The two time histories visually match, which can be observed in the bottom plot of Fig. 9(a) as a relatively straight line fits the two parameters. To examine the temperature

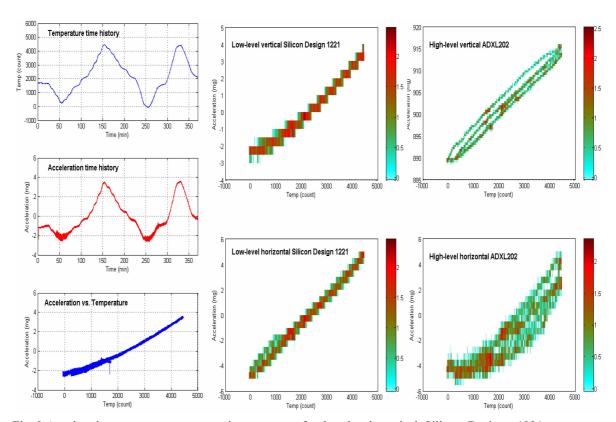


Fig. 9 Acceleration versus temperature in oven test for low-level vertical Silicon Designs 1221 sensor on prototype board: (a) time series, (b) two-dimensional histograms for Silicon Designs sensors and (c) two-dimensional histograms for ADXL sensors

sensitivity in more detail, Figs. 9(b) and 9(c) present two-dimensional histograms of the relationship between acceleration and temperature. In these figures, the concentration of points in any bin out of approximately 20,000 bins in each plot is logarithmically represented by the color codes.

The temperature tests demonstrate that the acceleration response generally varies linearly with temperature. For the Silicon Designs 1221 accelerometers on a typical sensor board (Board 1, Fig. 9(b)), this variation is 8 mg/45°C = 0.18 mg/°C. For ADXL 202 sensors on boards 1 in Fig. 9(c) the response is more non-linear and there is hysteresis behavior in the temperature response. Although for the channel in the horizontal direction the response variation is fairly small (about 8 mg), the channel in the vertical direction shows 25 mg response variation for the 45 degree temperature change and the difference between the hysteresis loops (loop diameter) is as large as 5 mg.

The hysteresis behavior of the sensors is not limited to ADXL 202 sensors; the response of Silicon Designs 1221 sensors on other sensor boards, subjected to the same temperature profile showed similar hysteresis loops. The hysteresis behavior of the response of some sensors to the temperature change is not fully understood. The oven test was repeated several times with different sensor boards, and similar results were observed. Further study of this phenomenon is left to future investigation. One possible explanation for this hysteresis behavior is the fact that the electronic circuitry inside the sensors generates small amount of heat and thus experience heating/cooling cycles independently of

the outside temperature. This cyclic heating pattern adds to the outside temperature change and affects the acceleration measurement.

4. Large-scale deployment on Golden Gate Bridge

Fig. 10 shows Golden Gate Bridge at the entrance of the San Francisco Bay, which has a 1280 m (4200 ft) long main-span and 343 m (1125 ft) side-spans. Two stiffening trusses support an orthotropic roadway deck and horizontal planes of wind bracing system at the bottom plane of the truss chords. The legs of the towers, 210 m (745 ft) above the water level, have cellular box sections, connected by horizontal struts at seven elevations (Strauss 1937, Stahl *et al.* 2007). The wireless sensor network that was designed and developed in this paper was deployed on the bridge to measure and record ambient accelerations. The sensor network consisted of 64 nodes on the main-span and the south tower of the bridge, as shown in Figs. 10 to 12. The network was designed to be scalable in terms of the number of the nodes, complexity of the network topology, data quality and quantity by addressing integrated hardware and software systems such as sensitivity and range of (MEMS) sensors, communication bandwidth of the low-power radio, reliability of command dissemination and data transfer, management of large volume of data and high-frequency sampling (Pakzad *et al.* 2008). The nodes on the main-span measure acceleration in vertical and transverse directions (Pakzad and Fenves 2009). On the tower, the nodes measure acceleration in transverse and longitudinal directions.

The instrumentation plan for the wireless sensor network for the bridge is shown in Fig. 10. The nodes on the main-span were located based on the range of the radio transmission distance at 30.5 m (100 ft) spacing, but a 15.25 m (50 ft) spacing was used where an obstruction hindered radio communication. Each main-span node was attached to the top flange of the floor girder directly inside of the cable. The eight nodes on the south tower were placed at the ends of four struts above

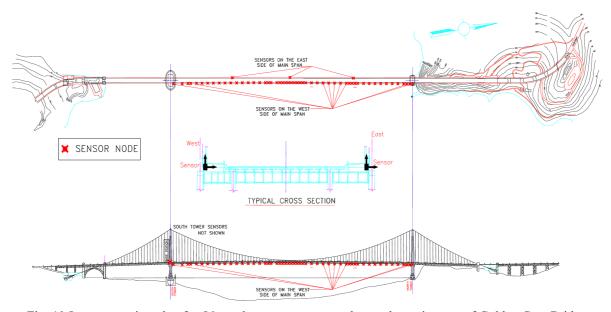


Fig. 10 Instrumentation plan for 56 accelerometer sensor nodes on the main-span of Golden Gate Bridge

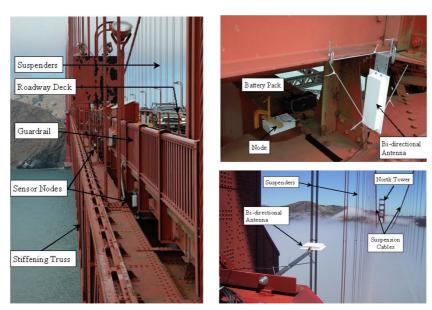


Fig. 11 Views of the main-span and south tower of Golden Gate Bridge as well as a node with its battery pack and bi-directional antenna installed on the main-span

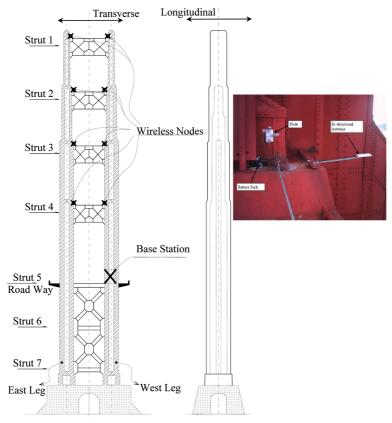


Fig. 12 Elevations of the south tower of Golden Gate Bridge, and temporary installation of a wireless sensor node on the tower

the roadway. Fig. 12 shows the elevation of the south tower and maps the location of the nodes. It also shows the temporary installation of a node on the tower with the antenna and battery pack. The tower nodes have a clear line of sight between them and hence have greater radio range than the main-span nodes. The node on the west side of the strut above the superstructure collects data from all the nodes on the tower and transmits them to the network on the main-span. Fifty-three (53) nodes were installed beginning on July 10, 2006, on the west side of the main-span. On September 15, 2006, the batteries were replaced for the nodes on the main-span and three extra nodes were added on the east side. The east side nodes were located at the two quarter-spans and the mid-span of the bridge and had radio communication with the west side nodes under the roadway deck. There were a total of 174 data collection runs of the network during the deployment which lasted until October 14, 2006, including testing and debugging, so all of the collected data sets do not contain data from all of the nodes.

The sampling rate for all runs was 1 kHz, but since the significant vibration frequencies of the bridge are much lower, the data were averaged on the node and downsampled to 50 to 200 Hz prior to transmission. The averaging is very effective in reducing the noise level and improving the accuracy of the estimated parameters. In some of the runs all five channels on a node (two highlevel motion sensors, two low-level motion sensors and the temperature sensor) were sampled, but in other runs the channels were limited to the low-level accelerometers to reduce the volume of data. The 512 kB flash memory of each node can buffer 250,000 samples of data, which may be allocated to any combination of the five sensor channels on the node (four accelerometer channels and a temperature sensor). Each run started with a pause to synchronize the network and disseminate a command to start sampling at a future time. After the scheduled sampling took place, there was a pause to establish the network routing. The recorded data were then transferred from each node to the base station using the reliable data communication and pipelining. Each run generated up to 500 kB data per node, which for the network of 60 nodes produced 30 MB data for 15 million samples. Approximately 1.3 GB data was collected during the deployment of the wireless sensor network on Golden Gate Bridge.

5. Statistical modal identification for the south tower

The ambient vibration data from the main-span of Golden Gate Bridge described in the previous section was examined and the result was presented in Pakzad and Fenves (2009). In this section the results of a statistical analysis of the identified vibration modes of the south tower is presented. Although the bridge including the south tower has been instrumented in the past, none of the earlier studies have provided data collected over an extended period of time to allow a statistical analysis of the modal properties. The contribution of this section is to demonstrate that the dense temporal sensing possible with WSNs, provides a high resolution and confidence in the identified vibration modes. The advantage of temporal density of the data is examined by estimating statistical properties for the identified vibration modes. The distribution of the identified mode shapes is examined by their confidence intervals.

Auto-Regressive with Moving Average (ARMA) method is used to identify the modal properties of the tower. For more information about the method as well as the choices for parameters refer to Pakzad and Fenves (2009), Pandit (1991), De Roeck *et al.* (1995), Andersen (1997), Heylen *et al.* (1995), and Pappa *et al.* (1993). For the mode shapes, the statistics are also presented by the mean

value of the parameters and their confidence intervals (CI). A 95% CI for a point-estimated parameter can be interpreted as an interval that is believed, with 95% confidence, to include the true value of the parameter. In other words, if the same procedures are repeated (sampling from the population, estimating the parameter and finding CIs), 95% of the times the estimated CIs are expected to include the true value of the parameter.

The statistical analysis allows inference about the certainty of the estimation of modal vibration properties. The narrower the confidence intervals are, the less uncertainty the estimated values have. The approach can also be used as a comparison basis for other estimations of the same modal parameters. A new estimate that lies inside the CI is consistent with the hypothesis that no change has occurred. If it falls outside the interval, the inconsistency can be explained by a change in the underlying parameter.

The results of system identification from all 174 runs are used to estimate statistical properties of vibration frequencies, damping ratios and mode shapes for the longitudinal, torsional and transverse modes of the tower.

The identified mode shapes, presented in Figs. 13-15 are generally consistent with the dynamic properties of the bridge tower. Note that in all cases since no sensors were installed on the tower below the roadway level, only the behavior of the upper part of the tower is included. The modal displacements are set to zero at the roadway level and the mode shapes are plotted relative to that. In the longitudinal direction the first mode frequency is 0.583 Hz, and the mode shape has no modal nodes (Fig. 13). The second, third and fourth frequencies are 1.805 Hz, 3.345 Hz and 4.733 Hz and the mode shapes have one, two and three modal nodes respectively. The damping ratios for these four modes are less than 0.7%. The first three torsional frequencies are 0.926 Hz, 2.239 Hz and 4.262 Hz, with damping ratios up to 1.5%, and zero, one and two modal nodes respectively. In the transverse direction the first mode frequency is 0.239 Hz, very close to the first transverse frequency of the main span. The estimated modal ordinates of this mode are very close to unity,

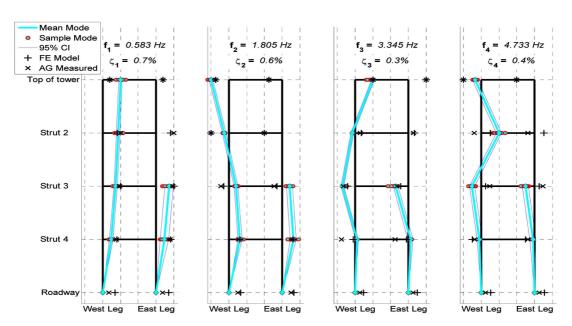


Fig. 13 Comparison between longitudinal modes of vibration of the south tower

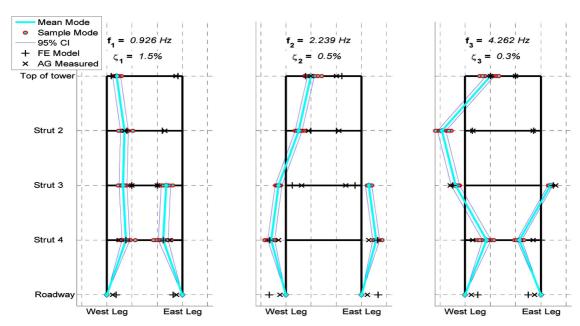


Fig. 14 Comparison between torsional modes of vibration of the south tower

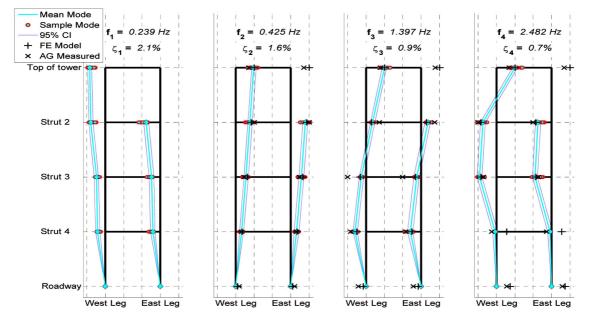


Fig. 15 Comparison between transverse modes of vibration of the south tower

which suggests that this mode is in fact the first transverse mode of the entire bridge. The second mode at 0.425 Hz frequency has a linear shape consistent with the first mode of a tower structure. The third and fourth modes at 1.397 Hz and 2.482 Hz have one and two modal nodes respectively.

The figures show the 95% confidence intervals for the mode shapes. Also included, are the results from an earlier deployments of accelerometers (Abdel-Ghaffar and Scanlan 1985) and finite element

models of the bridge (Abdel-Ghaffar *et al.* 1985). Note that in both of these cases the mode shapes are evaluated at all seven struts of the tower, from strut 1 at the top of the tower to strut 7 at the water level. The results in nearly all cases generally matches the CIs, with the cases where the modal ordinates at the roadway are closer to zero being more accurate than the others.

6. Conclusions

Development and large scale deployment of a MEMS-based wireless sensor network on a long span bridge was described. The hardware of the sensor nodes consists of a sensing unit of high- and low-level accelerometers and thermometers, control and filtering unit for on-board analog and digital data processing, and radio chips for wireless communication. Extensive laboratory testing was performed to validate the hardware system and evaluate its performance under different excitations and environmental circumstances. Each sensor board was individually calibrated using a tilt mechanism and the linearity of the response of each sensing channel was confirmed. Several prototype tests were performed to characterize the static and dynamic noise levels of the sensors. Low-level Silicon Designs has a RMS noise level of $71 \,\mu g/\sqrt{Hz}$ in the very low frequency range (below 0.1 Hz) which linearly reduces to $10 \,\mu g/\sqrt{Hz}$ for larger frequencies. The accelerometers performed well in shaking table dynamic tests with sweeping frequencies and matched the response of a reference sensor. All of the tested accelerometers showed some hysteresis response to changing temperature. Although the exact source of this behavior is not fully understood, but the micro heating of the accelerometer as a result of its operation is a contributing factor. Additional testing is required to investigate this phenomenon and address it in the future sensor design.

A large-scale wireless sensor network, consisting of 64 sensor nodes was deployed on Golden Gate Bridge for a period of more than three months and 174 sets of data were collected. The data consisted of a combination of ambient accelerations from low-level and high-level accelerometers and temperature. The statistical modal analysis of the vibration data for the south tower is presented in this paper. Longitudinal, transverse and torsional modal properties of the tower and their statistics are estimated. The mode shapes and their 95% confidence intervals are presented and compared with the estimated modes using finite element models as well as an earlier deployment of sensors on the bridge. The results in most cases match, in that the confidence intervals include the point estimates of the previous methods. It is also observed that the confidence intervals are very small, which indicates high quality and consistency of the data.

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